
Low-Cost Bidirectional Brushed DC Motor Control Using PIC16F684

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INTRODUCTION

This application note describes how to use the Enhanced, Capture, Compare, PWM (ECCP) on PIC16F684 for bidirectional, brushed DC (BDC) motor control. Low-cost BDC motor control can be used in applications such as intelligent toys, small appliances and power tools. PIC16F684 takes Microchip's mid-range family of products to the next level with its new ECCP peripheral. The ECCP peripheral builds on the technology of the CCP module with added features such as four PWM channels for easy bidirectional motor control through the hardware. This application note focuses on full-bridge configuration using the ECCP in PWM mode. The ECCP allows easy interfacing to a full-bridge configuration for bidirectional BDC motor control.

This application note describes the following:

- Calculating ECCP PWM Parameters
- Initializing the ECCP in Full-Bridge PWM mode
- Bidirectional BDC Motor Control
- Sensorless Motor Control Feedback
- Example Application

Note: All equations referenced in this application note can be found in [Appendix A](#).

CALCULATING ECCP PWM PARAMETERS

The PWM frequency, duty cycle and resolution need to be calculated when working with the ECCP in PWM mode.

Frequency

Selecting a PWM frequency for the motor control application affects the sound of the motor and the power transistor's switching speed. The human ear can detect frequencies ranging from 20 Hz to 20 kHz. In this application note, the PWM for motor control operates at 4 kHz. This results in less noise than even lower PWM frequencies, such as 1 kHz. If the application bandwidth can support higher frequencies (especially those above the range of typical human hearing), the motor will produce less audible noise. If too much audible noise is heard, the PWM frequency should be increased.

The PWM period and frequency can be calculated using Equations [A-1](#) and [A-2](#).

Duty Cycle

Changing the PWM duty cycle will change the average voltage across the motor, which in turn changes the motor's speed. The PWM duty cycle is calculated by using [Equation A-3](#). The average voltage across the BDC motor is calculated using [Equation A-4](#).

Resolution

The PWM duty cycle resolution determines the amount of precision with which the duty cycle can be changed. For example, a 10-bit resolution allows 1024 possible values for the duty cycle, where an 8-bit resolution only allows 256 values. The PWM frequency, the PIC16F684 oscillator frequency and Timer2 prescaler all affect the resolution value. The maximum resolution is ten bits. The PWM duty cycle resolution is calculated by using [Equation A-5](#).

INITIALIZING THE ECCP IN FULL-BRIDGE PWM MODE

When initializing the ECCP in Full-Bridge PWM mode, four registers need to be initialized:

PR2

The PR2 register affects the PWM frequency/period. The value to use for the PR2 register is calculated using [Equation A-6](#).

CCPR1L:CCP1CON<5:4>

The PWM duty cycle has a full resolution of ten bits. Since all registers on PIC16F684 are 8 bits wide, the ten bits are spread over two registers. CCPR1L contains the upper eight bits and CCP1CON<5:4> contains the lower two bits. The 10-bit value for CCPR1L:CCP1CON<5:4> is calculated using [Equation A-7](#).

CCP1CON

In addition to storing the lower two bits of the 10-bit PWM duty cycle, CCP1CON is used to set up the ECCP in PWM mode using bits CCP1CON<3:0>. It can also change the motor direction using bits CCP1CON<7:6>. When setting up the ECCP in PWM mode, there are four possible configurations. These configurations accommodate H-bridges with MOSFETS that are active-high, active-low or a combination of both active-high and active-low. Motor direction can be changed in hardware by configuring bits CCP1CON<7:6> to be '01' for forward or '11' for reverse. The PIC16F684 ECCP hardware switches channels for activating and modulating the appropriate MOSFET drivers in the H-bridge.

T2CON

The T2CON register is used for setting up the Timer2 prescaler and turning on Timer2. The Timer2 prescaler is contained in bits T2CON<1:0> and is used in determining the PWM frequency, duty cycle and resolution. Timer2 must be turned on by setting bit T2CON<2> before the PWM signal starts. An algorithm that calculates the Timer2 prescaler and PR2 values for PWM frequencies is shown in [Figure B-1](#).

BIDIRECTIONAL BDC MOTOR CONTROL

The ECCP makes changing the motor direction easy by configuring CCP1CON<7:6> to be '01' for forward (Figure 1) or '11' for reverse (Figure 2).

FIGURE 1: FULL-BRIDGE FORWARD CURRENT FLOW DIAGRAM

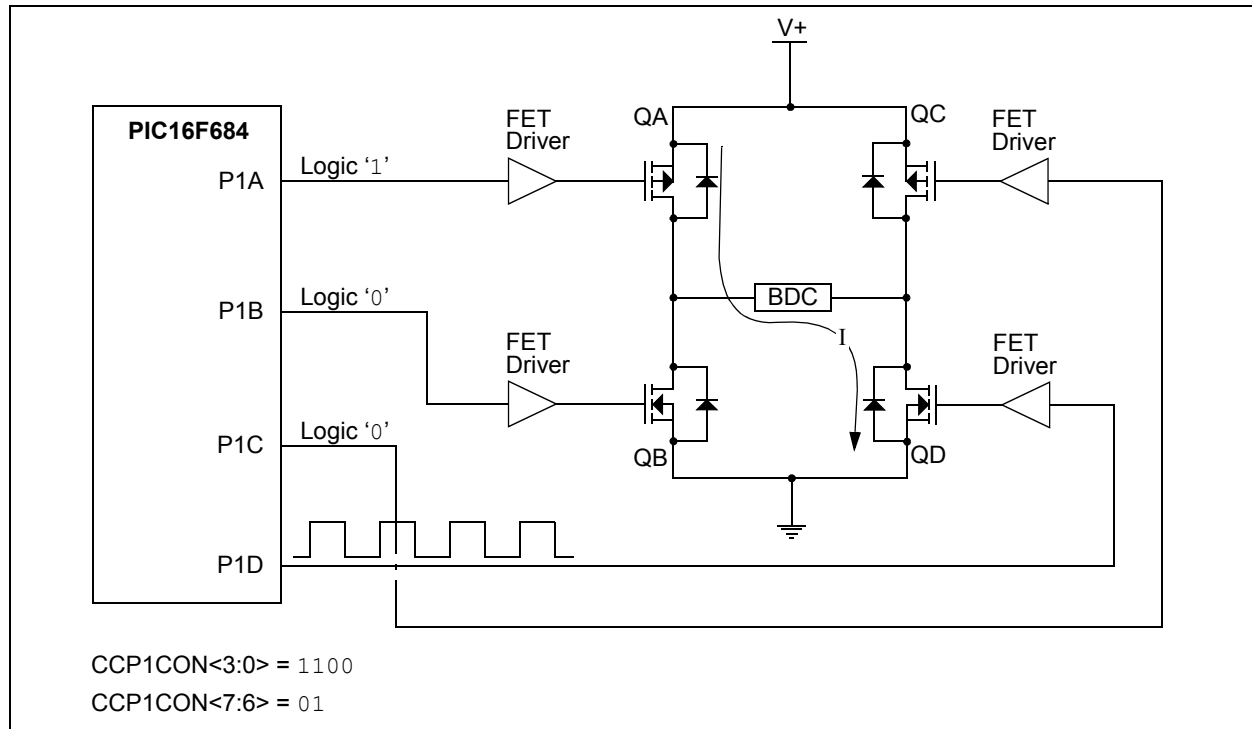
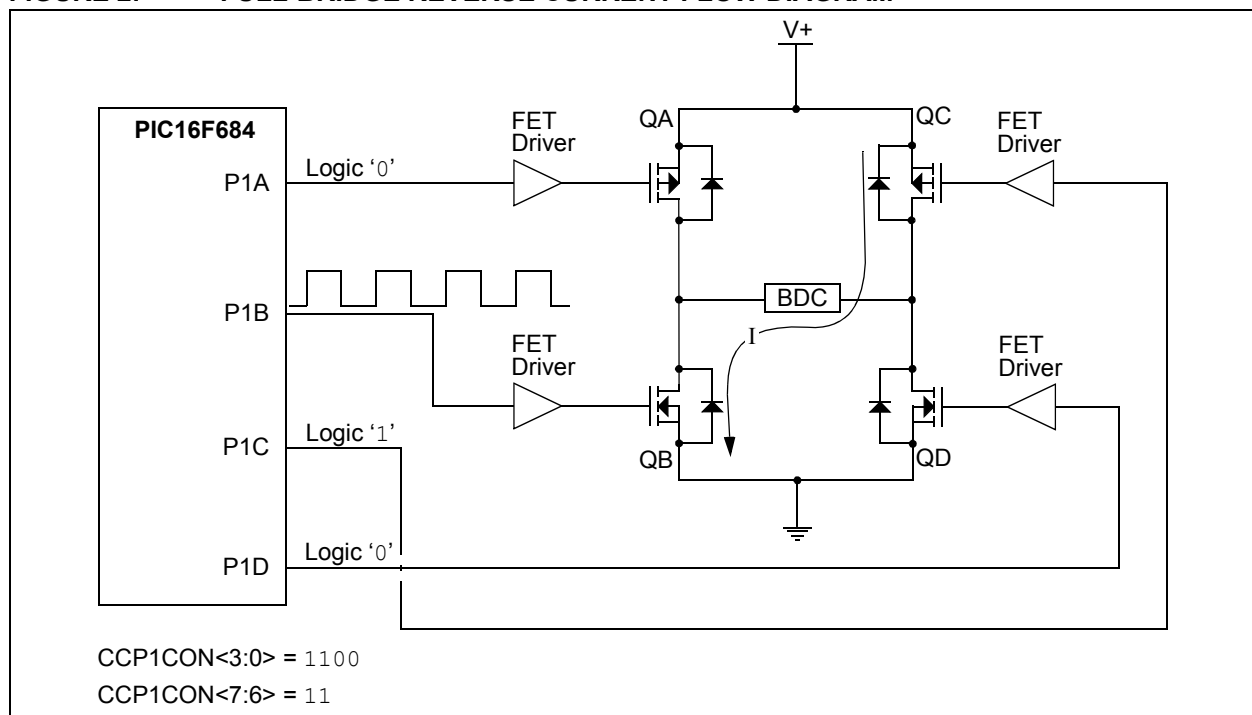


FIGURE 2: FULL-BRIDGE REVERSE CURRENT FLOW DIAGRAM



LOW COST SENSORLESS MOTOR CONTROL FEEDBACK

Sensorless RPM Measurement

Low-cost RPM measurement can be performed with a BDC motor by measuring the back EMF voltage from the motor (see [Figure 3](#)). The BDC RPM is directly proportional to the back EMF voltage. Since a BDC motor can be modeled as an inductive load, the voltage across the motor is equivalent to the inductance multiplied by di/dt . In this application, a 12V, 9600 max RPM BDC motor was used. To measure the back EMF voltage, turn off the modulated FET. This will cause the current to flow in the opposite direction. After initially shutting off the FET, di/dt must stabilize before taking the measurement. In order to use the PIC[®] microcontroller A/D converter, the measured voltage must be between 0V and V_{DD} . Since the back EMF voltage can be between 0V-12V, a voltage divider circuit is used to scale the back EMF voltage between 0V and V_{DD} . Using Microchip's MSP6S26 Programmable Gain Amplifier (PGA), a gain of 1 is used for buffering the scaled voltage that is being measured by the PIC16F684 A/D channel (see [Equation A-8](#) for calculating RPM).

Sensorless Current Measurement

Low-cost current measurements can be performed by using a current sensing resistor between the MOSFETS and ground (see [Figure 4](#)). To select a value appropriate for the resistance, consider the maximum amount of current allowed to flow through the resistor and the maximum amount of power dissipation.

In this application, a 0.1 ohm, 1W current sensing resistor was used with a maximum current of 3A. When 3A flow through the resistor, the ideal power dissipated in the resistor is 0.9W (see [Equation A-9](#)) and the voltage across the resistor is 0.3V (see [Equation A-10](#)). In order to get the most resolution from the 10-bit A/D converter, the voltage across the resistor at 3A must be amplified as close as possible to the PIC16F684 V_{DD} , which is 5V in this application. Using Microchip's MSP6S26 PGA, a gain of 16 will ideally give 4.8V, at the maximum 3A specified current (see [Equation A-11](#)). A gain of 16 gives a 9.94-bit A/D resolution for measuring current (see [Equations A-12](#) and [A-13](#)). The current through the resistor can then be computed using [Equations A-14](#), [A-15](#) and [A-16](#).

Since a PWM signal is used to drive the BDC motor, the H-bridge circuit only draws current during the high pulse-width of the PWM period. To obtain a current measurement, the voltage across the current sensing resistor is sampled over a PWM period. A sampling and averaging algorithm of taking measurements over multiple PWM periods is shown in [Figure B-2](#).

FIGURE 3: FULL-BRIDGE FORWARD CONFIGURATION WITH BACK EMF MEASUREMENT

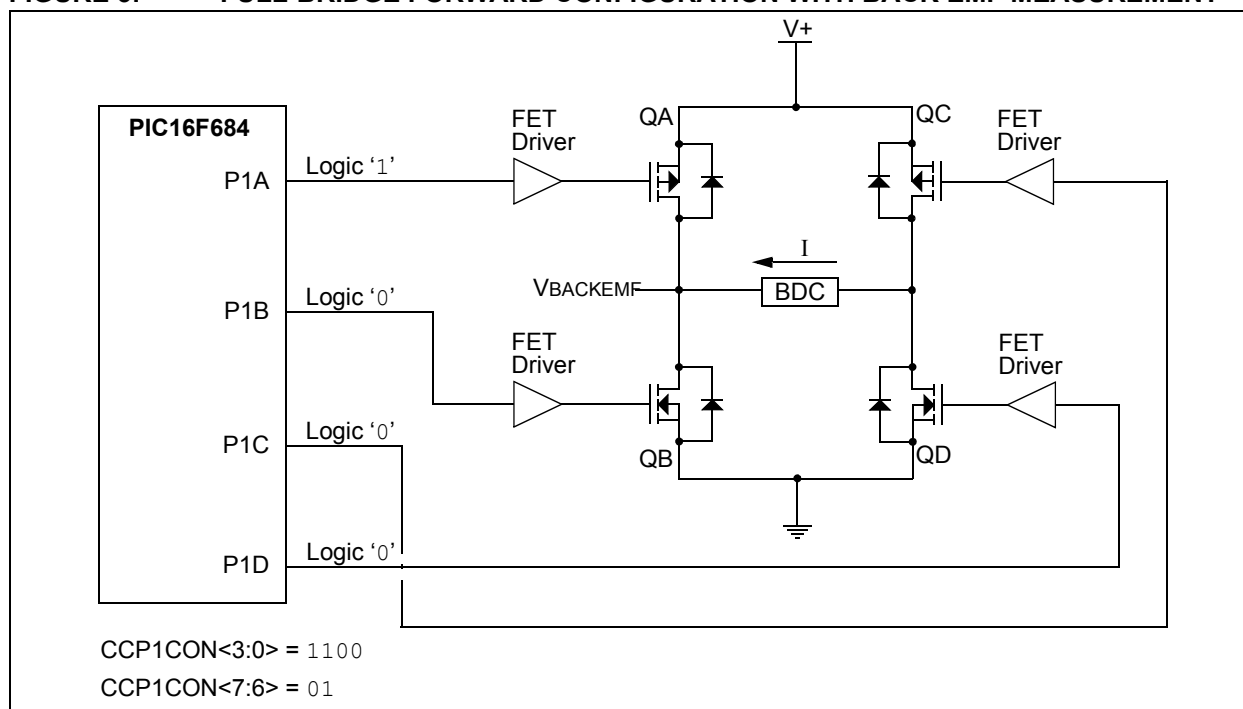
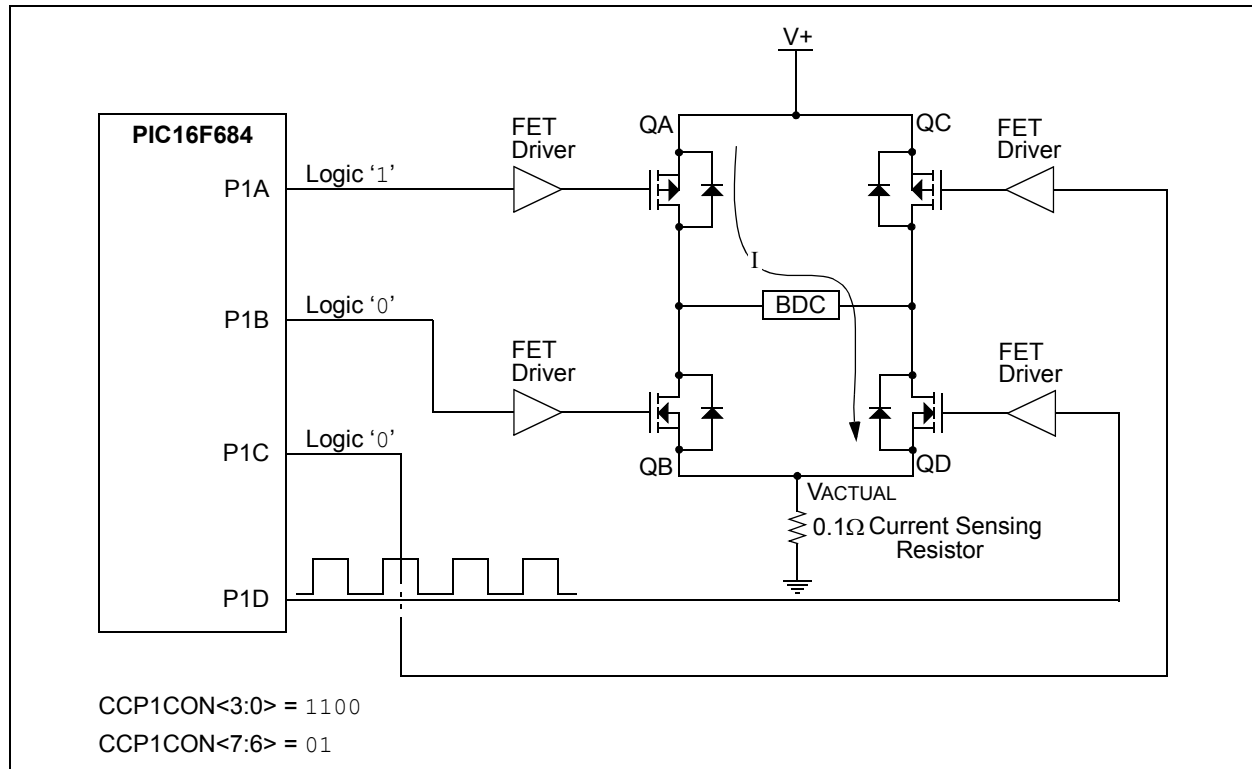
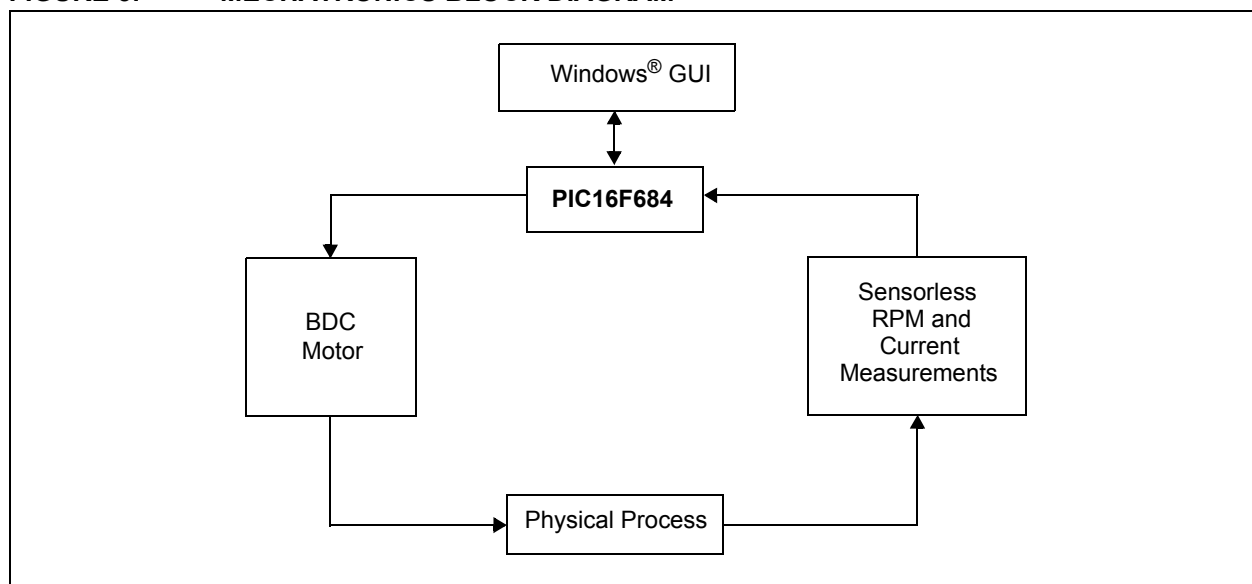


FIGURE 4: FULL-BRIDGE FORWARD WITH CURRENT-SENSING RESISTOR**EXAMPLE APPLICATION**

This example application demonstrates a low-cost BDC motor control system using the ECCP configured in Full-Bridge PWM mode (see Figure 5). The user interface allows the user to easily configure a BDC motor with PIC16F684, adjust the PWM frequency and

duty cycle, change the PIC16F684 internal oscillator frequency in real-time, and view RPM and current measurements. This application source code was written using the HI-TECH C[®] compiler, MPLAB[®] IDE, and the Microsoft Visual C++[®] 6.0 development platform.

FIGURE 5: MECHATRONICS BLOCK DIAGRAM

Firmware

The example firmware is responsible for many operations:

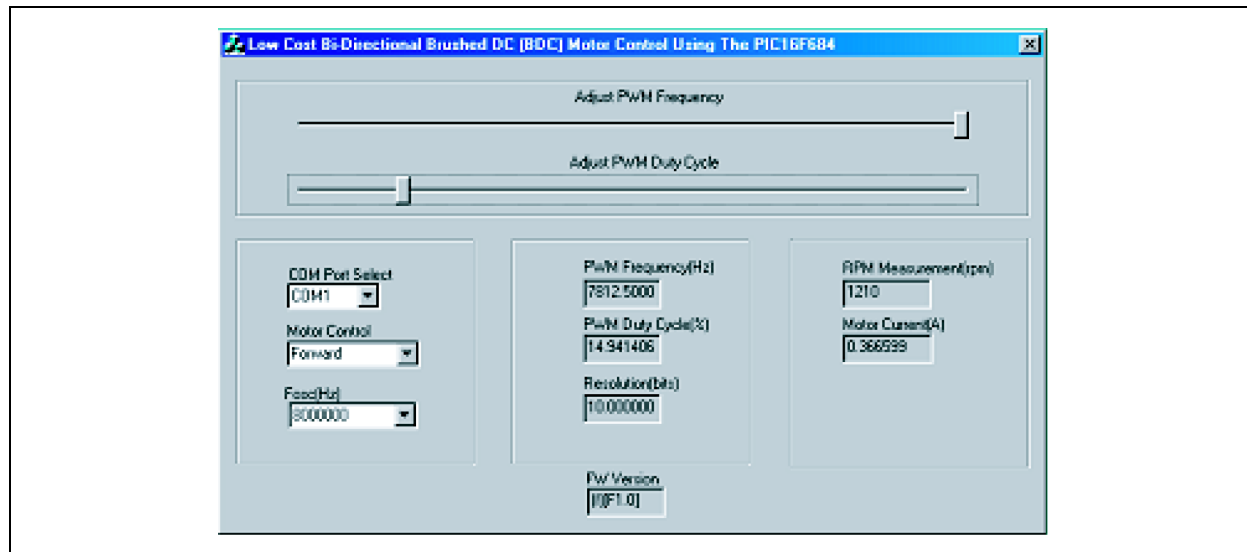
- Initializing the PIC16F684
- Sending bit-banged SPI commands to the PGA
- Receiving commands from the PC
- Modifying the PWM frequency and duty cycle
- Changing the motor's direction
- Changing the internal oscillator frequency
- Taking A/D converter measurements for RPM and current

The PIC16F684 firmware implements a bit-banged RS-232 USART running at 9600 bps. See [Appendix C](#) for the RS-232 serial protocol used in this application note. The C source code can be downloaded from www.microchip.com. See [Figure B-3](#) for the main program flow.

Software

The Windows® user interface provides the user a friendly environment for interfacing the BDC motor. The user interface allows the user to adjust the PWM frequency, duty cycle, motor direction and internal oscillator frequency. The user interface also displays the PWM frequency, duty cycle, resolution, RPM and current. The PC software is the host and sends commands to the PIC16F684 using RS-232. The Windows user interface source code can also be downloaded from www.microchip.com. The Windows user interface example is shown in [Figure 6](#).

FIGURE 6: WINDOWS® USER INTERFACE SCREEN



Hardware

The hardware used in this application note contains three major sections:

- Power stage for motor control
- Communication for RS-232
- Measurement for RPM and current

The power stage consists of a full H-bridge used for bidirectional BDC motor control. PIC16F684 uses RC2-RC5 as the four ECCP pins that interface with the full H-bridge circuit.

The communication section consists of an RS-232 serial communication configuration. PIC16F684 uses RA5 for sending and receiving RS-232 data.

The measurement section consists of Microchip's MSC6S26 multi-channel PGA and a voltage divider circuit for scaling the back EMF voltage, as discussed in [Section "Sensorless RPM Measurement"](#).

PIC16F684 communicates to the PGA via a 3-wire bit-banged SPI interface. The CS pin is connected to RA1. The SCK pin is connected to RA2. The SI pin is connected to RC0. The VREF pin is connected to GND. The RA0 pin is used as an analog input for measuring RPM and current. The RA0 pin is connected to the VOUT pin on the PGA. Channel 0 on the PGA is used for RPM measurements. Channel 1 on the PGA is used for current measurements. See [Figure D-1](#) for the schematic diagram of the hardware.

CONCLUSION

PIC16F684 is well suited for low-cost bidirectional BDC motor control. This application note demonstrates how easy it is to calculate the necessary parameters for using the ECCP in PWM mode, to initialize the necessary ECCP registers, use the ECCP for bidirectional BDC motor control and implement sensorless RPM and current measurements. This application note concludes by showing a full application implementation using PC Windows software, PIC16F684 firmware and motor control hardware.

REFERENCES

1. *PIC16F684 14-Pin, Flash-Based 8-Bit CMOS Microcontrollers with nanoWatt Technology* Data Sheet (DS41202): www.microchip.com/PIC16F684
2. *MCP6S21/2/6/8 Single-Ended, Rail-to-Rail I/O, Low Gain PGA* Data Sheet (DS21117): www.microchip.com/MCP6S26
3. MPLAB® IDE: www.microchip.com/archives

APPENDIX A:EQUATIONS

EQUATION A-1: PWM FREQUENCY (HZ)

$$Frequency = \frac{1}{Period}$$

EQUATION A-2: PWM PERIOD (SECONDS)

$$Period = [(PR2 + 1)] \times 4 \times TOSC \times TMR2Prescaler$$

EQUATION A-3: DUTY CYCLE (SECONDS)

$$DC = CCPR1L:CCP1CON<5:4> \times TOSC \times TMR2Prescaler$$

EQUATION A-4: VOLTAGE ACROSS BDC MOTOR (VOLTS)

$$V_{BDC} = V_{DD} \times \left(\frac{DC}{Period} \right)$$

EQUATION A-5: RESOLUTION (BITS)

$$Resolution = \frac{\log\left(\frac{FOSC}{(FPWM \times TMR2Prescaler)}\right)}{\log(2)}$$

EQUATION A-6: PR2

$$PR2 = \left(\frac{Period}{4 \times TOSC \times TMR2Prescaler} \right) - 1$$

EQUATION A-7: CCPR1L:CCP1CON<5:4>

$$CCPR1L:CCP1CON<5:4> = \frac{DC}{TOSC \times TMR2Prescaler}$$

EQUATION A-8: RPM

$$RPM = \left(1 - \left(\frac{ADRESH:ADRESL}{1024} \right) \right) \times RPM_{MAX}$$

EQUATION A-9: POWER (W)

$$P = I_{MAX}^2 \times R = 3^2 \times 0.1 = 0.9W$$

EQUATION A-10: MAXIMUM VOLTAGE ACROSS RESISTOR (VOLTS)

$$V_{NOMINALMAX} = I_{MAX} \times R = 3 \times 0.1 = 0.3V$$

EQUATION A-11: MAXIMUM VOLTAGE AFTER AMPLIFICATION (VOLTS)

$$V_{GAINMAX} = V_{NOMINALMAX} \times Gain = 0.3 \times 16 = 4.8V$$

EQUATION A-12: BITS OF RESOLUTION

$$2^X = \frac{V_{GAINMAX}}{V_{DD}} \times 1024, \text{ where } X \text{ is bits of resolution}$$

EQUATION A-13: BITS OF RESOLUTION SOLVED FOR X

$$X = \log \frac{\left(\frac{V_{GAINMAX}}{V_{DD}} \times 1024 \right)}{\log(2)} = \frac{\log \left(\frac{4.8}{5.0} \times 1024 \right)}{\log(2)} = 9.94 \text{ bits}$$

EQUATION A-14: GAIN VOLTAGE MEASURED (VOLTS)

$$V_{GAIN} = \left(\frac{ADRESH:ADRESL}{2^X} \right) \times V_{GAINMAX}$$

EQUATION A-15: ACTUAL VOLTAGE ACROSS RESISTOR (VOLTS)

$$V_{ACTUAL} = \frac{V_{GAIN}}{Gain}$$

EQUATION A-16: CURRENT THROUGH RESISTOR (VOLTS)

$$I = \frac{V_{ACTUAL}}{R}$$

APPENDIX B: FLOWCHARTS

FIGURE B-1: CALCULATING TIMER2 PRESCALER AND PR2 ALGORITHM GIVEN A PWM FREQUENCY

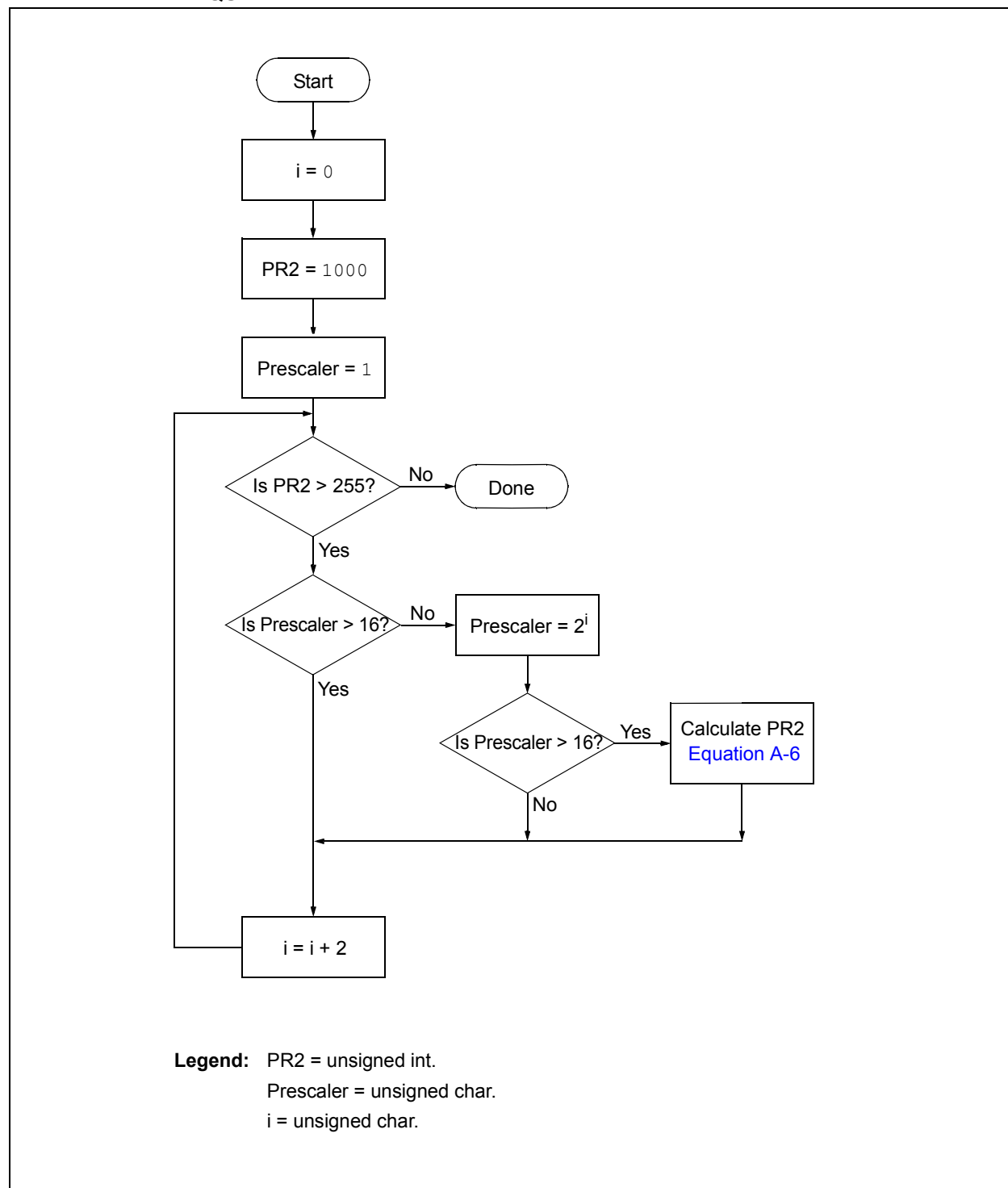
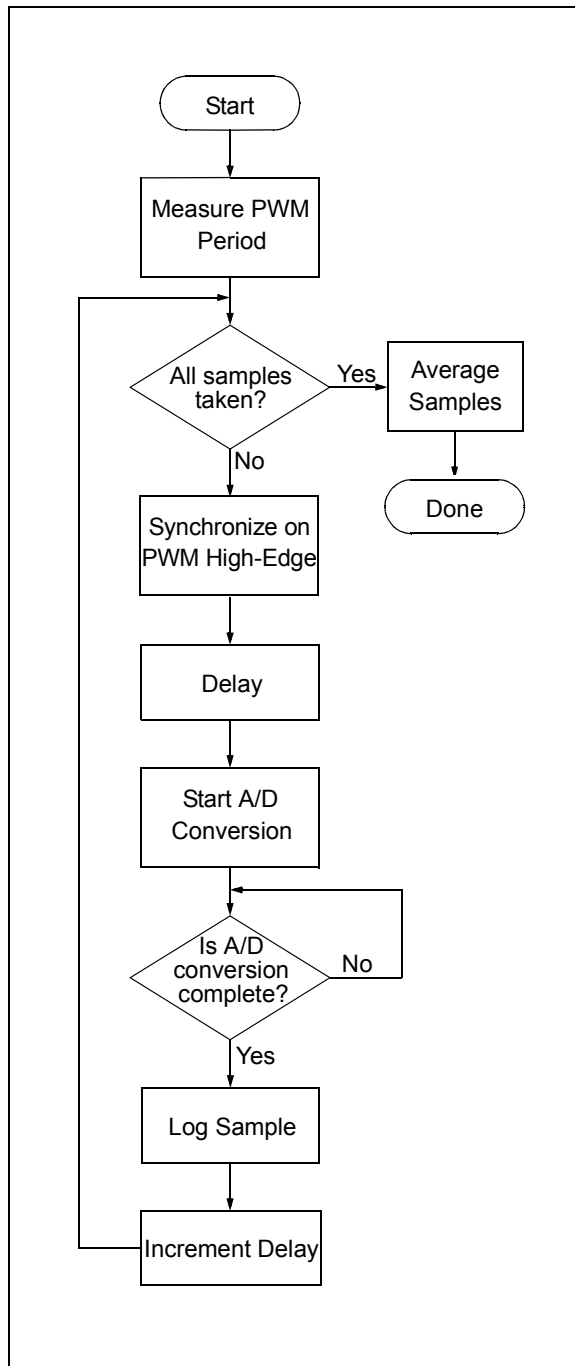
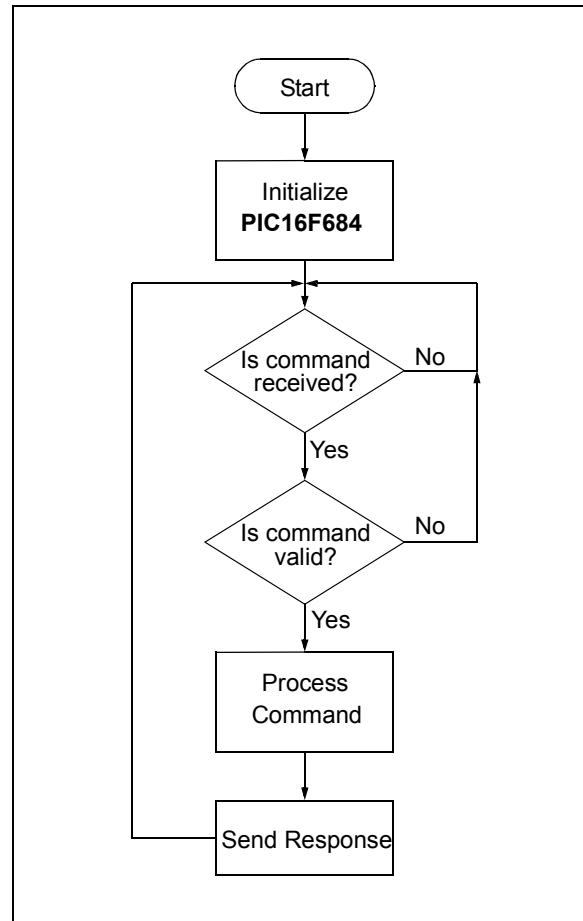


FIGURE B-2: PWM SAMPLING AND AVERAGING ALGORITHM**FIGURE B-3: MAIN ROUTINE**

APPENDIX C: RS-232 SERIAL COMMUNICATIONS PROTOCOL

Since one-wire communication is being implemented, the command sent from the PC to PIC16F684 will be echoed back. An example of this can be seen on the firmware version box in the Windows GUI. The firmware version box contains (f) [F1.0]. The PC command sent is (f). The PIC16F684 firmware response is [F1.0]. The general form of the command and response are described below as well as the commands implemented in the example application.

C.1 General Form

PC Command:

`<command start><command><data> <command end>`

Ex: (f).

PIC16F684 Response:

`<response start><response><data><response end>`

Ex: [F1.0]

Note 1: The `<command>` is lower case.

2: The `<response>` is the upper case of the `<command>`.

3: If there is no `<data>` to be sent, the `<command end>` can be the next character sent.

4: All `<data>` is sent in Hex format.

5: All `<data>` is sent Most Significant Byte first.

6: Invalid commands are ignored and responded with a [?].

7: Invalid `<command start>` is ignored and not responded to.

8: Commands and responses are currently set to ten characters each, this can be adjusted in the source code on both the Windows software and PIC16F684 firmware.

C.2 Example Application Command Set

PR2 Command: Loads data into the PR2 register.

PC Command: (aAF)

PIC16F684 Response: [A]

CCPR1L Command: Loads data into the CCPR1L register.

PC Command: (b1F)

PIC16F684 Response: [B]

CCP1CON<5:4> Command: Loads data into CCP1CON<5:4>.

PC Command: (c3)

PIC16F684 Response: [C]

Timer2 Prescaler Command: Loads data into T2CON<1:0>.

PC Command: (d0)

PIC16F684 Response: [D]

Fosc Command: Loads data into OSCCON<6:4>.

PC Command: (e6)

PIC16F684 Response: [E]

FW Command: Requests the PIC16F684 firmware version.

PC Command: (f)

PIC16F684 Response: [F1.0]

Motor Control Command: Loads data into CCP1CON<7:6>.

PC Command: (g3)

PIC16F684 Response: [G]

RPM Measurement Command: Requests a RPM measurement.

PC Command: (h)

PIC16F684 Response: [H3FF]

Current Measurement Command: Requests a Current measurement.

PC Command: (i)

PIC16F684 Response: [I2BC]

APPENDIX D: SCHEMATICS

FIGURE D-1: HARDWARE SCHEMATIC

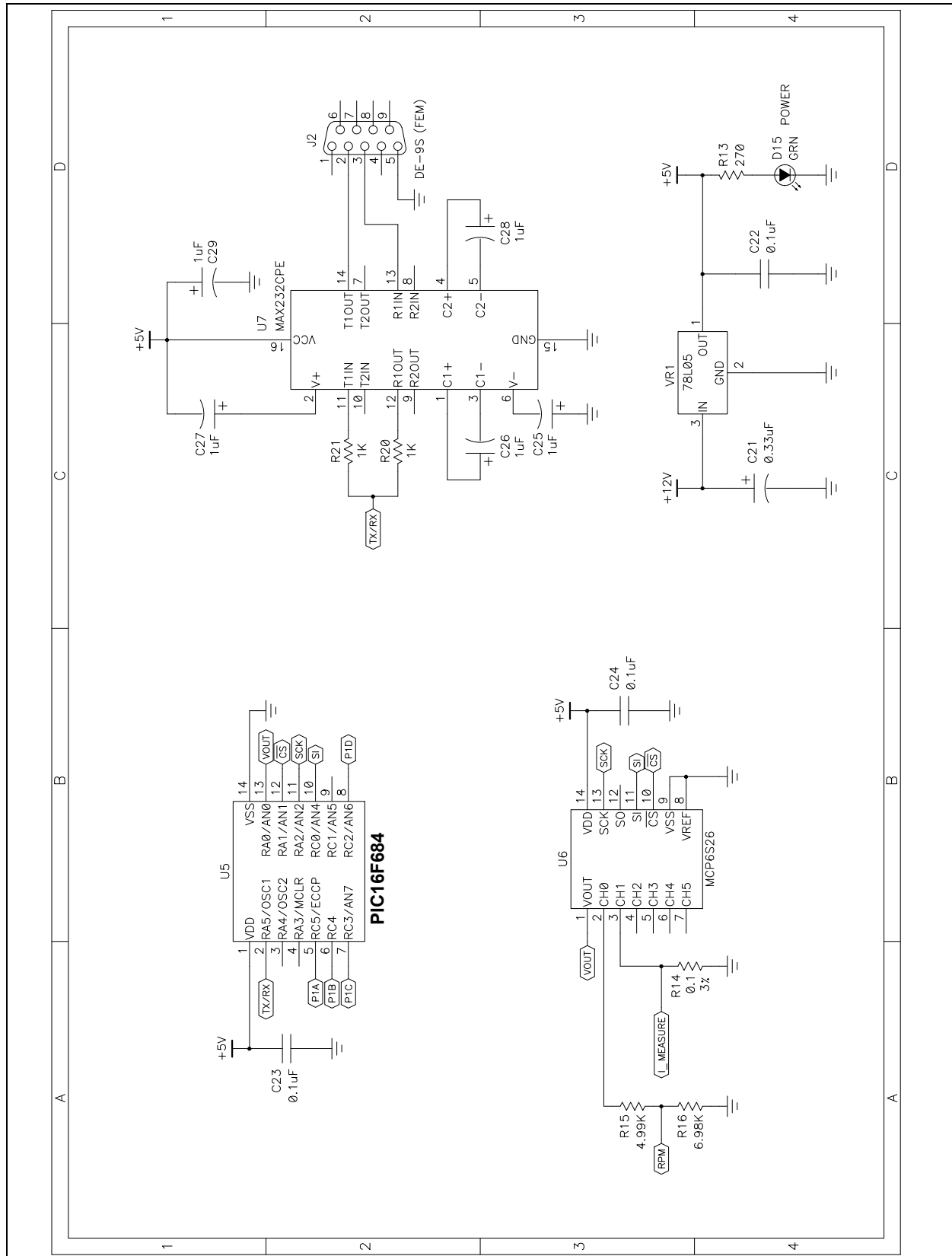
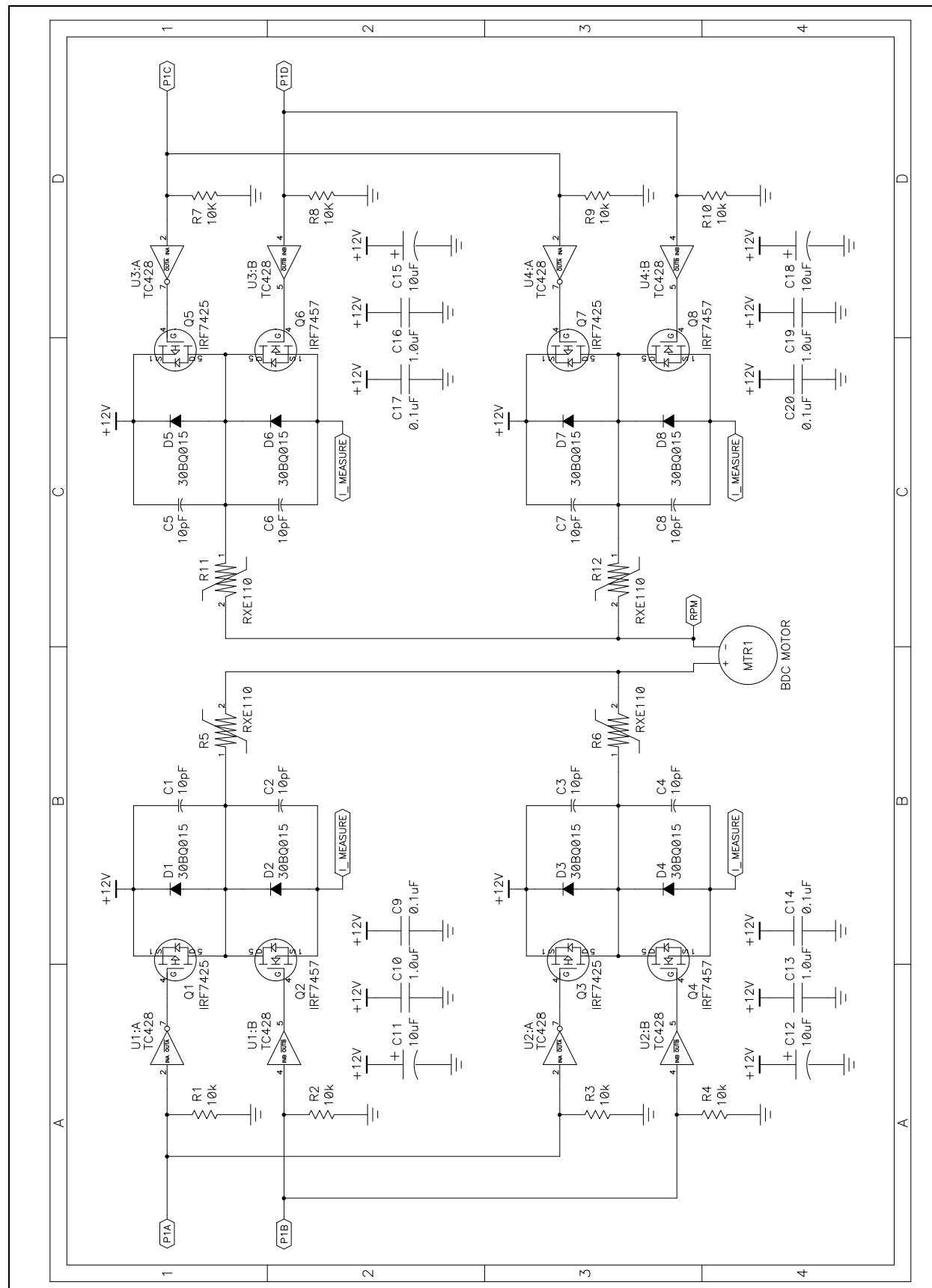


FIGURE D-2: BDC MOTOR CONTROL SCHEMATIC



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